

# PHYSICS 169

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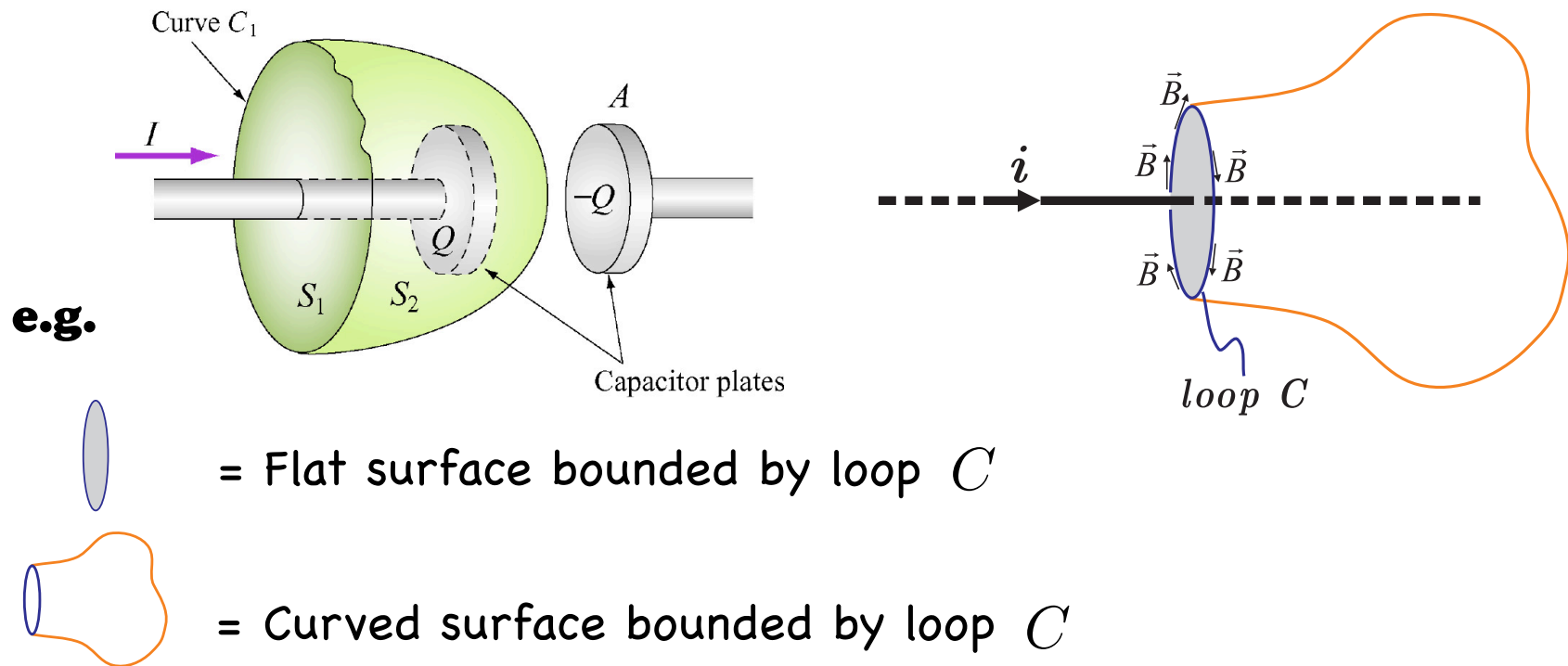
Tuesday, April 24, 18

## 11.1 Displacement Current

We can use **Ampere's law** to calculate magnetic fields due to currents

Integral  $\oint_C \vec{B} \cdot d\vec{s}$  around any close loop  $C$  is equal to  $\mu_0 i_{incl}$

where  $i_{incl} =$  **current passing an area bounded by closed curve  $C$**



If **Ampere's law** is true all time

then  $i_{incl}$  **determined should be independent of surface chosen**


Consider a simple case: charging a capacitor

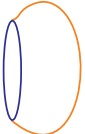
We know there is a current flowing  $i(t) = \frac{\mathcal{E}}{R} e^{-t/RC}$

which leads to a magnetic field observed  $\vec{B}$

With **Ampere's law**  $\oint_C \vec{B} \cdot d\vec{s} = \mu_0 i_{\text{inc}}$

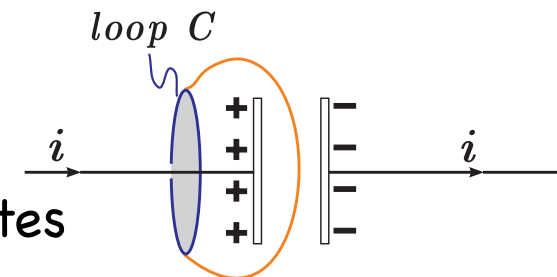
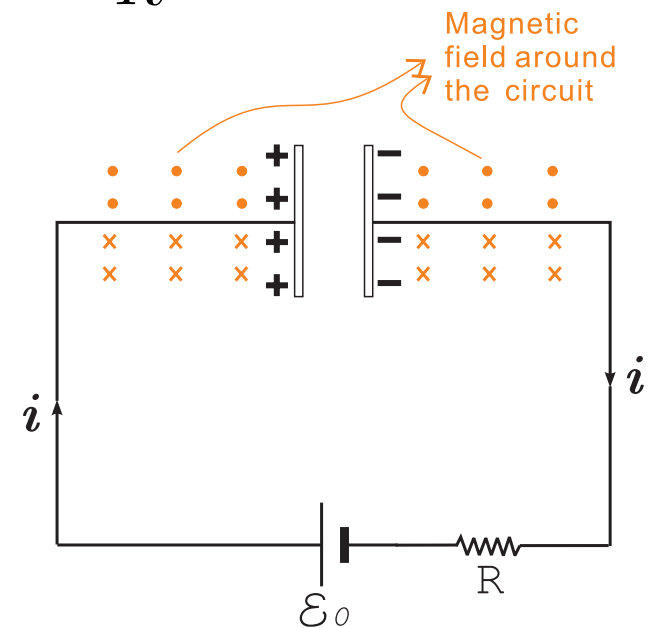
BUT WHAT IS  $i_{\text{inc}}$  ?

If we look at   $i_{\text{inc}} = i(t)$

If we look at   $i_{\text{inc}} = 0$

$\therefore$  There is no charge flow between capacitor plates

$\therefore$  Ampere's law is either WRONG or INCOMPLETE





Two observations

① While there is no current between capacitor's plates there is a **time-varying electric field between plates of capacitor**

② We know **Ampere's law is mostly correct from measurements of B-field around circuits**



**Can we revise Ampere's law to fix it?**

Electric field between capacitor's plates  $E = \frac{\sigma}{\epsilon_0} = \frac{Q}{\epsilon_0 A}$

$Q$  = charge on capacitor's plates

$A$  = area of capacitor's plates

$$\therefore Q = \epsilon_0 \underbrace{E \cdot A}_{\text{Electric flux}} = \epsilon_0 \Phi_E$$



∴ We can define

$$\frac{dQ}{dt} = \epsilon_0 \frac{d\Phi_E}{dt} = i_{\text{dis}}$$

where  $i_{\text{dis}}$  is called **Displacement Current** (first proposed by Maxwell)

Maxwell first proposed that this is missing term for Ampere's law

$$\oint_C \vec{B} \cdot d\vec{s} = \mu_0 \left( i_{\text{inc}} + \epsilon_0 \frac{d\Phi_E}{dt} \right) \quad \textbf{Ampere-Maxwell law}$$

$i_{\text{inc}}$  = current through any surface bounded by  $C$

$\Phi_E$  = electric flux through that **same surface bounded by curve  $C$**

$$\Phi_E = \int_S \vec{E} \cdot d\vec{a}$$

## 11.2 Induced Magnetic Field

Electric field can be generated by

☞ charges

☞ changing magnetic flux

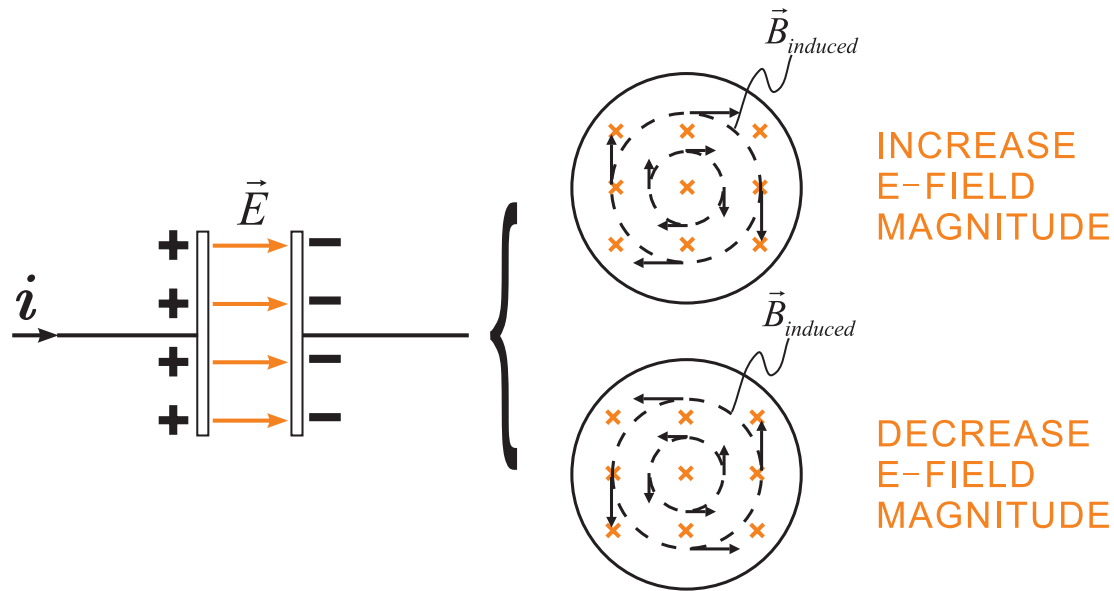
Ampere-Maxwell law that a magnetic field can be generated by

☞ moving charges (current)

☞ changing electric flux

A change in electric flux through a surface bounded by  $C$

can lead to an induced magnetic field along loop  $C$



**Notes** Induced magnetic field is along **same direction** as caused by **changing electric flux**

**Example** What is magnetic field strength inside a circular plate capacitor of radius  $R$  with a current  $I(t)$  charging it?

**Answer** Electric field of capacitor

$$E = \frac{Q}{\epsilon_0 A} = \frac{Q}{\epsilon_0 \pi R^2}$$

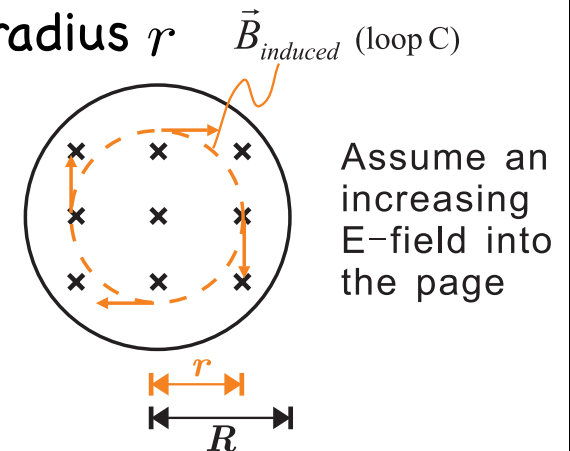
Electric flux inside capacitor through a loop  $C$  of radius  $r$

Ampere-Maxwell Law inside capacitor:

$$\underbrace{\oint_C \vec{B} \cdot d\vec{s}}_{\substack{\because \vec{B}_{\text{induced}} \parallel d\vec{s} \\ \text{Length of loop } C}} = \mu_0 \left( i_{\text{incl}} + \epsilon_0 \frac{d\Phi_E}{dt} \right)$$

$$2\pi r B_{\text{induced}} = \mu_0 \epsilon_0 \frac{d}{dt} \left( \frac{Q r^2}{\epsilon_0 R^2} \right)$$

$$= \mu_0 \frac{r^2}{R^2} \underbrace{\frac{dQ}{dt}}_{I(t)}$$



$$\therefore B_{\text{induced}} = \frac{\mu_0 r}{2\pi R^2} I(t) \quad \text{for } r < R$$



Outside capacitor plate  $\blacktriangleright$

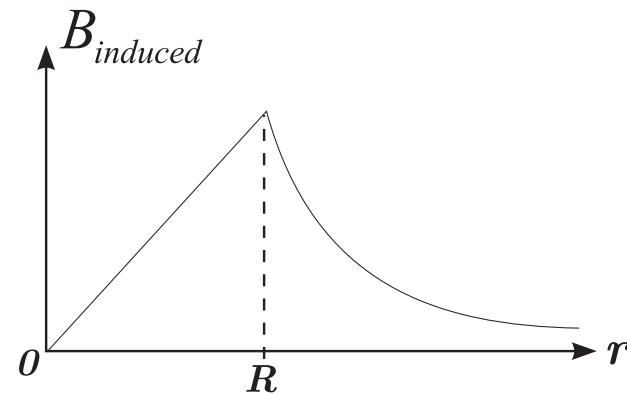
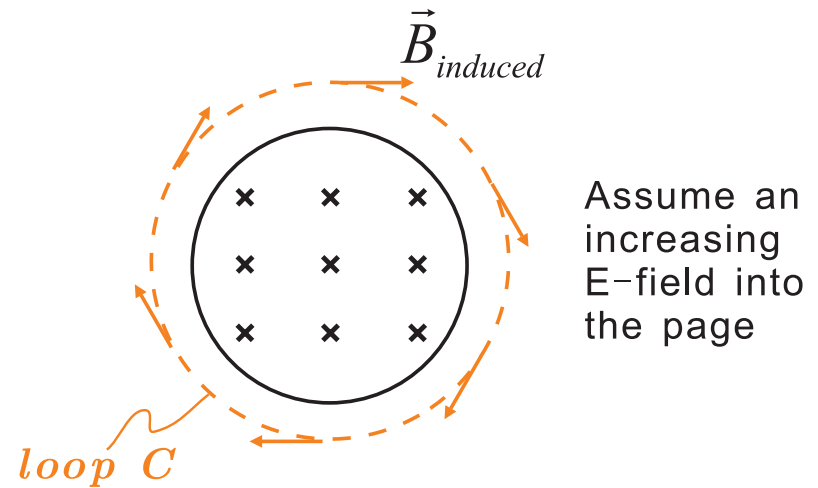
Electric flux through loop  $C$

$$\Phi_E = E \cdot \pi R^2 = \frac{Q}{\epsilon_0}$$

$$\oint_C \vec{B} \cdot d\vec{s} = \mu_0 \left( i_{\text{inc}} + \epsilon_0 \frac{d\Phi_E}{dt} \right)$$

$$2\pi r B_{\text{induced}} = \mu_0 \epsilon_0 \left( \frac{1}{\epsilon_0} \cdot \frac{dQ}{dt} \right)$$

$$\therefore B_{\text{induced}} = \frac{\mu_0 I(t)}{2\pi r}$$



## 11.3 Maxwell's Equations

Four equations that **completely** describe ↗

behaviors of electric and magnetic fields

$$\oint_S \vec{E} \cdot d\vec{A} = \frac{Q_{\text{inc}}}{\epsilon_0} = \frac{1}{\epsilon_0} \int_V \rho dV$$

$$\oint_S \vec{B} \cdot d\vec{A} = 0$$

$$\oint_C \vec{E} \cdot d\vec{s} = -\frac{d}{dt} \int_S \vec{B} \cdot d\vec{A}$$

$$\oint_C \vec{B} \cdot d\vec{s} = \mu_0 i_{\text{inc}} + \mu_0 \epsilon_0 \frac{d}{dt} \int_S \vec{E} \cdot d\vec{A}$$

One equation that describes ↗

**how matter reacts to electric and magnetic fields**

$$\vec{F} = q(\vec{E} + \vec{v} \times \vec{B})$$

## Maxwell's equations in matter

$$\oint_S \vec{D} \cdot d\vec{A} = \int_V \rho \, dV$$

$$\oint_{\partial\Omega} \vec{B} \cdot d\vec{A} = 0$$

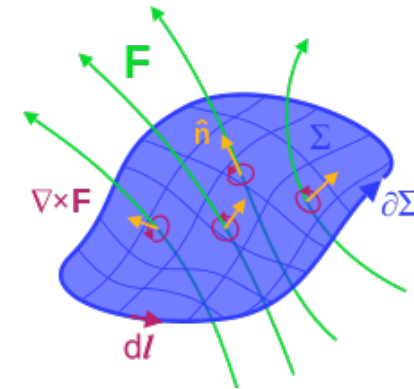
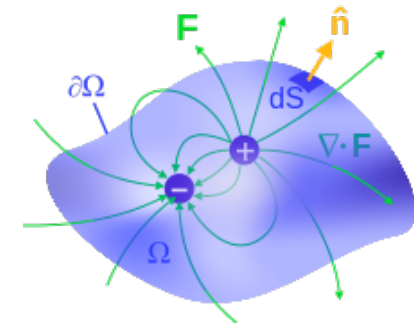
$$\oint_C \vec{E} \cdot d\vec{s} = -\frac{d}{dt} \int_S \vec{B} \cdot d\vec{A}$$

$$\oint_C \vec{H} \cdot d\vec{s} = \int_S \vec{J} \cdot d\vec{A} + \frac{d}{dt} \int_S \vec{D} \cdot d\vec{A}$$

Electric displacement field  $\Rightarrow \vec{D} = \epsilon_0 \vec{E} + \vec{P}$

vacuum  $\Rightarrow \vec{D} = \epsilon_0 \vec{E}$

isotropic linear dielectric  $\Rightarrow \vec{D} = \epsilon \vec{E}$



Boundary conditions  $\Rightarrow$

	Component	General materials	Linear materials
Electric displacement	Perpendicular	$D_{2,\perp} - D_{1,\perp} = \sigma_f$	$D_{2,\perp} - D_{1,\perp} = \sigma_f$
	Parallel	$D_{2,\parallel} - D_{1,\parallel} = ?$	$\frac{D_{2,\parallel}}{\epsilon_2} = \frac{D_{1,\parallel}}{\epsilon_1}$
Electric field	Perpendicular	$\epsilon_2 E_{2,\perp} - \epsilon_1 E_{1,\perp} = ?$	$\epsilon_2 E_{2,\perp} - \epsilon_1 E_{1,\perp} = \sigma_f$
	Parallel	$E_{2,\parallel} = E_{1,\parallel}$	$E_{2,\parallel} = E_{1,\parallel}$



Maxwell's equations may also be written in differential form

$$\nabla \cdot \mathbf{D} = \rho$$

$$\nabla \cdot \mathbf{B} = 0$$

$$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}$$

$$\nabla \times \mathbf{H} = \mathbf{J} + \frac{\partial \mathbf{D}}{\partial t}$$

Important consequence of Maxwell's equations

prediction electromagnetic waves that travel @ speed of light

Reason is due to the fact that

changing electric field produces a magnetic field and vice versa

Coupling between 2 fields leads to generation of electromagnetic waves

Prediction was confirmed by Hertz in 1887

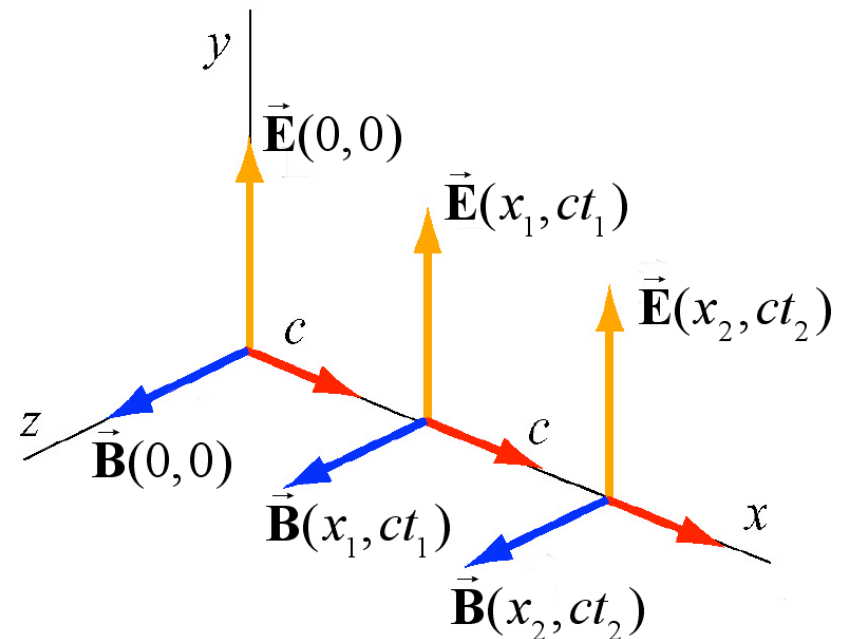
## 11.4 Plane Traveling Electromagnetic Waves

Consider electromagnetic wave propagating in  $+x$  direction with uniform  $\vec{E}$  pointing in  $+y$ -direction and  $\vec{B}$  in  $+z$ -direction

At any instant both  $\vec{E}$  and  $\vec{B}$  are independent of  $(y, z)$  coordinates

$$\vec{E}(x, t) = E_y(x, t)\hat{j}$$

$$\vec{B}(x, t) = B_z(x, t)\hat{k}$$



This (non-physical) electric and magnetic field yield a **plane wave** because at any instant both  $\vec{E}$  and  $\vec{B}$  are uniform over any plane perpendicular to direction of propagation

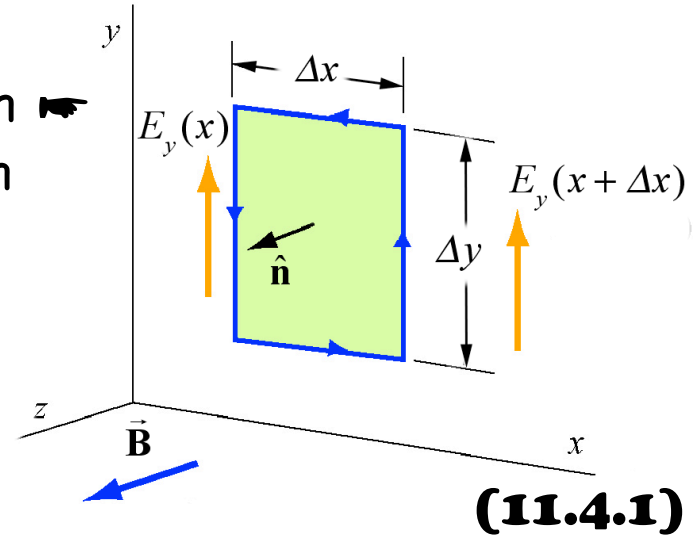
Consider a rectangular loop that lies in  $xy$ -plane  
 left side of loop at  $x$  and right at  $x + \Delta x$   
 bottom at  $y$  and top at  $y + \Delta y$  as shown in

Unit vector normal to loop positive  $z$ -direction

$$\hat{n} = \hat{k}$$

Recall Faraday's law

$$\oint \vec{E} \cdot d\vec{s} = -\frac{d}{dt} \int \int \vec{B} \cdot d\vec{A} \quad (11.4.1)$$



To evaluate LHS of (11.4.1) integrate around closed path

$$\oint \vec{E} \cdot d\vec{s} = E_y(x + \Delta x) \Delta y - E_y(x) \Delta y \quad (11.4.2)$$

use Taylor expansion to approximate

$$E_y(x + \Delta x) = E_y(x) + \frac{\partial E_y}{\partial x} \Delta x + \dots \quad (11.4.3)$$

Left-hand-side of Faraday's law becomes

$$\oint \vec{E} \cdot d\vec{s} = \frac{\partial E_y}{\partial x} \Delta x \Delta y \quad (11.4.4)$$



Assume that  $\Delta x$  and  $\Delta y$  are very small such that time derivative of  $z$ -component of magnetic field is nearly uniform over area element

Rate of change of magnetic flux on right-hand-side of Eq. (11.4.1) is

$$-\frac{d}{dt} \int \int \vec{B} \cdot d\vec{A} = -\frac{\partial B_z}{\partial t} \Delta x \Delta y \quad \textbf{(11.4.5)}$$

Equating two sides of Faraday's Law and dividing through by area  $\Delta x \Delta y$

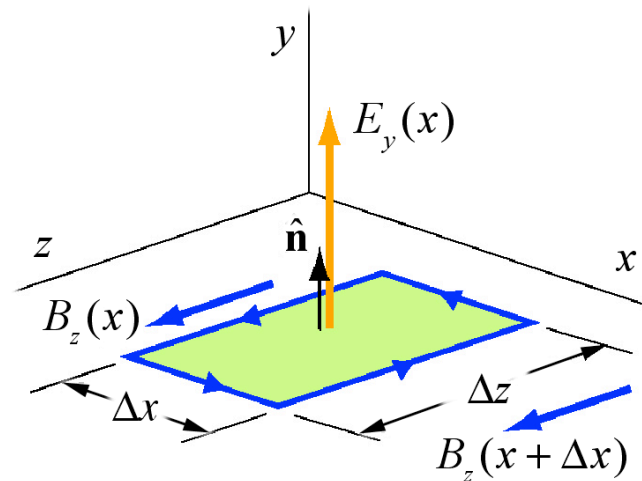
$$\frac{\partial E_y}{\partial x} = -\frac{\partial B_z}{\partial t} \quad \textbf{(11.4.6)}$$

Eq. (11.4.6) indicates that at each point in space time-  
varying B-field is associated with spatially varying E-field

Second condition on relationship between electric and magnetic fields may be deduced by using Ampere-Maxwell equation

$$\oint_C \vec{B} \cdot d\vec{s} = \mu_0 i_{\text{inc}} + \mu_0 \epsilon_0 \frac{d}{dt} \int_S \vec{E} \cdot \vec{A} \quad \textbf{(11.4.7)}$$

Consider a rectangular loop in  $xy$ - plane depicted



Evaluating line integral of magnetic field around closed path

$$\oint \vec{B} \cdot d\vec{s} = B_z(x) \Delta z - B_z(x + \Delta x) \Delta z \quad \textbf{(11.4.8)}$$

Use Taylor expansion to approximate

$$B_z(x + \Delta x) = B_z(x) + \frac{\partial B_z}{\partial x} \Delta x + \dots \quad \mathbf{(11.4.9)}$$

Left-hand-side of Maxwell-Ampere law becomes

$$\oint \vec{B} \cdot d\vec{s} = -\frac{\partial B_z}{\partial x} \Delta x \Delta z \quad \mathbf{(11.4.10)}$$

Assuming that  $\Delta x$  and  $\Delta z$  are very small such that time derivative of  $y$ -component of electric field is nearly uniform over area element  
Rate of change of electric flux on right-hand-side of Eq.(11.4.7) is

$$\mu_0 \epsilon_0 \frac{d}{dt} \int \int \vec{E} \cdot d\vec{A} = \mu_0 \epsilon_0 \frac{\partial E_y}{\partial t} \Delta x \Delta z \quad \mathbf{(11.4.11)}$$

Equating two sides of Maxwell-Ampere law and dividing by  $\Delta x \Delta z$  yields

$$-\frac{\partial B_z}{\partial x} = \mu_0 \epsilon_0 \frac{\partial E_y}{\partial t} \quad \mathbf{(11.4.12)}$$

Eq. (11.4.12) indicates that at each point in space  
time-varying E-field is associated to spatially varying B-field

**Eq. (11.4.6)** and **Eq. (11.4.12)** are coupled differential equations

To uncouple them

first take another partial derivative of **Eq. (11.4.6)** with respect to  $x$

$$\frac{\partial^2 E_y}{\partial x^2} = -\frac{\partial}{\partial x} \left( \frac{\partial B_z}{\partial t} \right) = -\frac{\partial}{\partial t} \left( \frac{\partial B_z}{\partial x} \right) \quad \text{Eq. (11.4.13)}$$

Assumed that field  $B_z$  is sufficiently well behaved  
such that partial derivatives are interchangeable

$$\frac{\partial}{\partial x} \left( \frac{\partial B_z}{\partial t} \right) = \frac{\partial}{\partial t} \left( \frac{\partial B_z}{\partial x} \right) \quad \text{Eq. (11.4.14)}$$

Substitute **Eq. (11.4.12)** into **Eq. (11.4.13)**

**One-dimensional wave equation**



$$\frac{\partial^2 E_y}{\partial x^2} = \mu_0 \epsilon_0 \frac{\partial^2 E_y}{\partial t^2} \quad \text{Eq. (11.4.15)}$$

By a dimensional analysis

quantity  $\frac{1}{\mu_0 \epsilon_0}$  has same dimensions as speed squared

Repeat argument to find a one-dimensional wave equation satisfied by  $z$ -component of magnetic field  $\rightarrow$  taking  $\partial/\partial x$  of **Eq. (11.4.12)**

$$-\frac{\partial^2 B_z}{\partial x^2} = \mu_0 \epsilon_0 \frac{\partial}{\partial x} \frac{\partial E_y}{\partial t} = \mu_0 \epsilon_0 \frac{\partial}{\partial t} \left( \frac{\partial E_y}{\partial x} \right) \quad \text{Eq. (11.4.16)}$$

Substitute **Eq. (11.4.6)** into **Eq. (11.4.16)** yielding a one-dimensional wave equation satisfied by  $z$ -component of magnetic field

$$\frac{\partial^2 B_z}{\partial x^2} = \mu_0 \epsilon_0 \frac{\partial^2 B_z}{\partial t^2} \quad \text{Eq. (11.4.17)}$$

General form of a one-dimensional wave equation is given by

$$\frac{\partial^2 \Psi(x, t)}{\partial x^2} = \frac{1}{v^2} \frac{\partial^2 \Psi(x, t)}{\partial t^2} \quad \text{Eq. (11.4.18)}$$

where  $v$  is **speed of propagation** and  $\Psi(x, t)$  is **wave function**

$E_y$  and  $B_z$  satisfy wave equation and propagate with speed

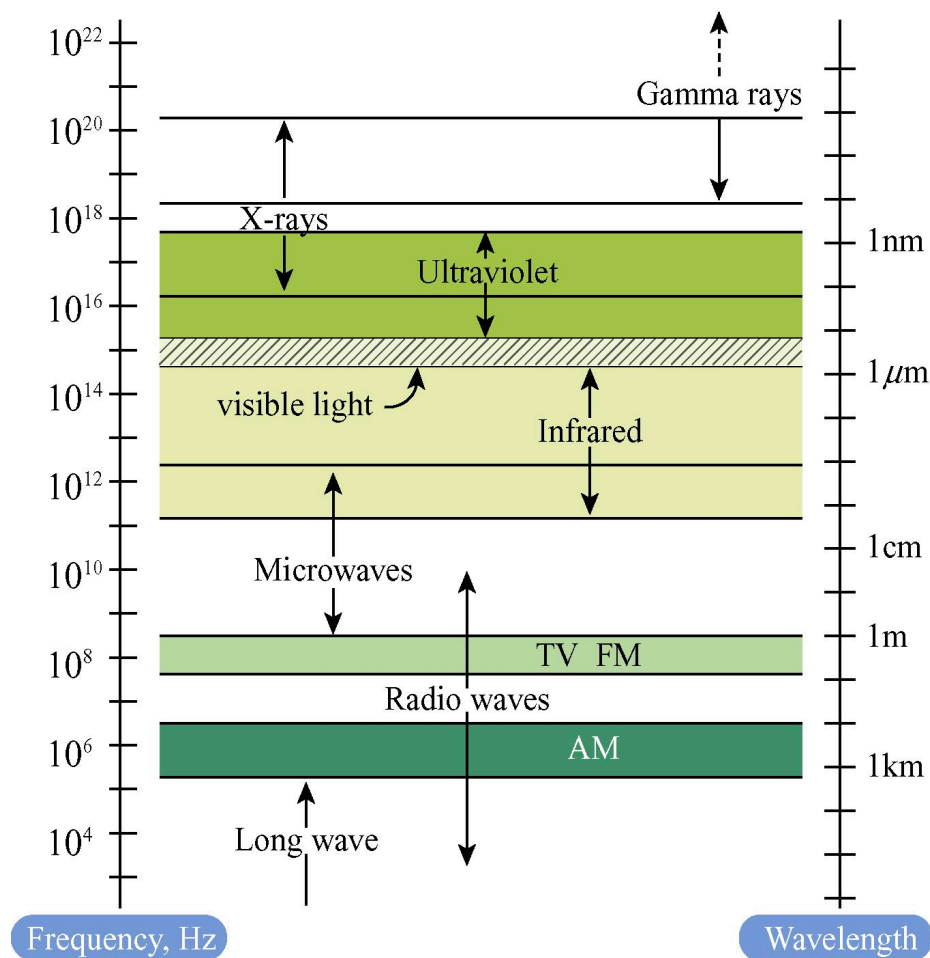
$$v = \frac{1}{\sqrt{\mu_0 \epsilon_0}} \quad \text{Eq. (11.4.19)}$$

Taking  $\mu_0 = 4\pi \times 10^{-7} \text{ Tm/A}$  and  $\epsilon_0 = 8.854 \times 10^{-12} \text{ C/N/m}^2$

$$v = c = 2.997 \times 10^8 \text{ m/s}$$

Maxwell's Equations predict that E- and B-fields propagate through space at speed of light

### Electromagnetic spectrum



And God Said

$$\nabla \cdot \mathbf{E} = \frac{\rho}{\epsilon_0}$$

$$\nabla \cdot \mathbf{B} = 0$$

$$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}$$


$$\nabla \times \mathbf{B} = \mu_0 \mathbf{J} + \mu_0 \epsilon_0 \frac{\partial \mathbf{E}}{\partial t}$$

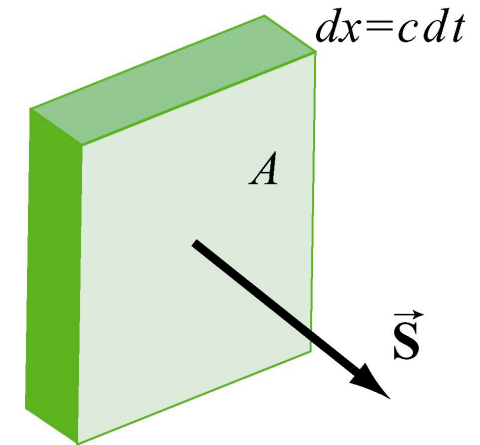
and then there was

"Light"

## 11.5 Poynting Vector

Energy can also be transported by electromagnetic waves

Consider plane EM wave passing through small volume element of area  $A$  and thickness  $dx$  as shown in 



Total energy stored in electromagnetic fields in volume element is



$$dU = u A dx = (u_E + u_B) A dx = \frac{1}{2} \left( \epsilon_0 E^2 + \frac{B^2}{\mu_0} \right) A dx \quad \text{Eq. (11.5.1)}$$

$$u_E = \frac{1}{2} \epsilon_0 E^2 \quad u_B = \frac{B^2}{2\mu_0} \quad \text{Eq. (11.5.2)}$$

Because electromagnetic wave propagates with speed of light  $c$  time it takes for wave to move through volume element is  $dt = dx/c$

One may obtain rate of change of energy per unit area

$$S = \frac{dU}{A dt} = \frac{c}{2} \left( \epsilon_0 E^2 + \frac{B^2}{\mu_0} \right) \quad \text{Eq. (11.5.3)}$$



$SI$  unit of  $S$  is  $[W \cdot m^{-2}]$

Recall that magnitude of fields satisfy  $E = cB$  and  $c = 1/\sqrt{\mu_0\epsilon_0}$

Therefore **Eq. (11.5.3)** may be rewritten as

$$S = \frac{c}{2} \left( \epsilon_0 E^2 + \frac{B^2}{\mu_0} \right) = \frac{cB^2}{\mu_0} = c\epsilon_0 E^2 = \frac{EB}{\mu_0} \quad \mathbf{Eq. (11.5.4)}$$

Turn this energy flow into a vector

by assigning direction as direction of propagation

Rate of energy flow per unit area is called **Poynting vector**  $\vec{S}$

(after British physicist John Poynting) and defined by vector product

$$\vec{S} = \frac{1}{\mu_0} \vec{E} \times \vec{B} \quad \mathbf{Eq. (11.5.5)}$$

Plane transverse electromagnetic waves

fields  $\vec{E}$  and  $\vec{B}$  are perpendicular and magnitude of  $\vec{S}$

$$|\vec{S}| = \frac{\vec{E} \times \vec{B}}{\mu_0} = \frac{EB}{\mu_0} = S \quad \mathbf{Eq. (11.5.6)}$$

As an example, suppose electric field associated with a plane sinusoidal electromagnetic wave is  $\vec{E} = E_0 \cos(kx - \omega t) \hat{j}$

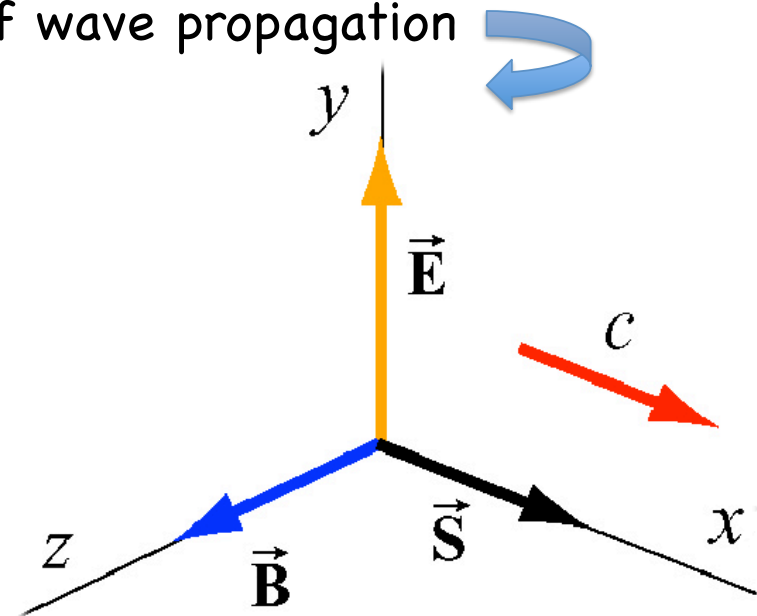
Corresponding magnetic field is  $\vec{B} = B_0 \cos(kx - \omega t) \hat{k}$

and direction of propagation is positive  $x$ -direction

Poynting vector is then  $\blacktriangleright$  **Eq. (11.5.7)**

$$\vec{S} = \frac{1}{\mu_0} (E_0 \cos(kx - \omega t) \hat{j}) \times (B_0 \cos(kx - \omega t) \hat{k}) = \frac{E_0 B_0}{\mu_0} \cos^2(kx - \omega t) \hat{i}$$

As expected  $\blacktriangleright$   $\vec{S}$  points in direction of wave propagation



**Intensity** of wave  $I$  is defined as time-average of  $S$

$$I = \langle S \rangle = \frac{E_0 B_0}{\mu_0} \langle \cos^2(kx - \omega t) \rangle = \frac{E_0 B_0}{2\mu_0} = \frac{E_0^2}{2c\mu_0} = \frac{cB_0^2}{2\mu_0} \quad \text{Eq. (11.5.8)}$$

$$\text{recall } \langle \cos^2(kx - \omega t) \rangle = \frac{1}{2} \quad \text{Eq. (11.5.9)}$$

To relate intensity to energy density

we first note equality between electric and magnetic energy densities

$$u_B = \frac{B^2}{2\mu_0} = \frac{(E/c)^2}{2\mu_0} = \frac{E^2}{2c^2\mu_0} = \frac{\epsilon_0 E^2}{2} = u_E \quad \text{Eq. (11.5.10)}$$

Time-averaged energy density of wave is then

$$\langle u \rangle = \langle u_E + u_B \rangle = \epsilon_0 \langle E^2 \rangle = \frac{\epsilon_0}{2} E_0^2 = \frac{1}{\mu_0} \langle B^2 \rangle = \frac{B_0^2}{2\mu_0} \quad \text{Eq. (11.5.11)}$$

Comparing **Eqs. (11.5.8)** and **Eq. (11.5.11)**

we can conclude that intensity is related to average energy density by

$$I = \langle S \rangle = c \langle u \rangle \quad \text{Eq. (11.5.12)}$$

## 11.6 Momentum and Radiation Pressure

An electromagnetic wave transports not only energy but also momentum and hence can exert a radiation pressure on a surface due to absorption and reflection of momentum

When a plane electromagnetic wave is completely absorbed by a surface momentum transferred is related to energy absorbed by

$$\Delta p = \frac{\Delta U}{c} \quad \text{complete absorption} \quad \text{Eq. (11.6.1)}$$

(We shall not prove this result as it involves a more complicated description of energy and momentum stored in electromagnetic fields)

If EM wave is completely reflected by a surface such as a mirror



$$\Delta p = \frac{2\Delta U}{c} \quad \text{complete reflection} \quad \text{Eq. (11.6.2)}$$

For a wave that is completely absorbed  
time-averaged radiation pressure (force per unit area) is given by

$$P = \frac{\langle F \rangle}{A} = \frac{1}{A} \left\langle \frac{dp}{dt} \right\rangle = \frac{1}{Ac} \left\langle \frac{dU}{dt} \right\rangle \quad \text{Eq. (11.6.3)}$$

Because time-averaged rate that energy delivered to surface is

$$\left\langle \frac{dU}{dt} \right\rangle = \langle S \rangle A \quad \text{Eq. (11.6.4)}$$

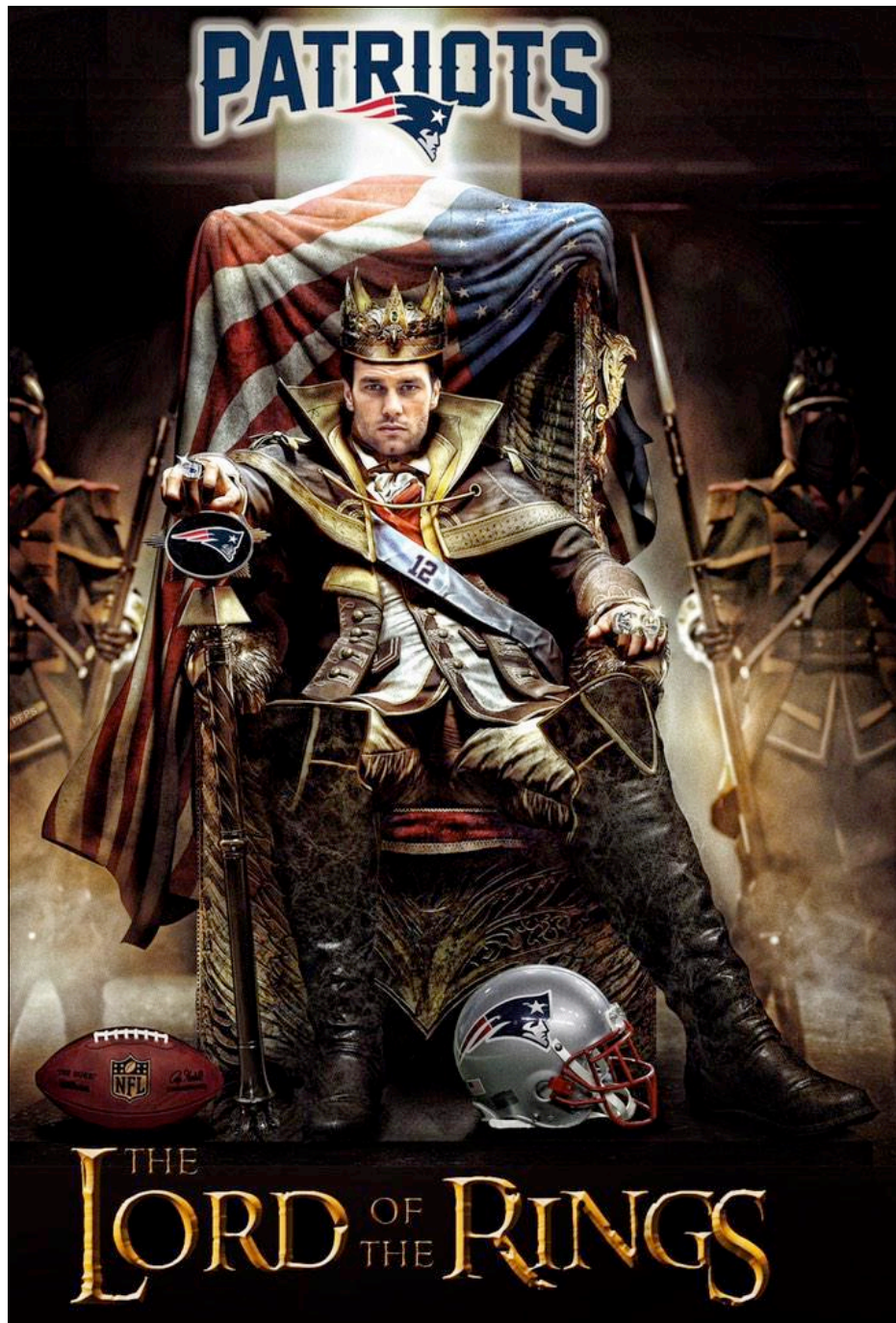
Substitute **Eq. (11.6.4)** into **Eq. (11.6.3)** yielding

$$P = \frac{\langle S \rangle}{c} \quad \text{complete absorption} \quad \text{Eq. (11.6.5)}$$

If radiation is completely reflected  
radiation pressure is twice as great as case of complete absorption

$$P = \frac{2\langle S \rangle}{c} \quad \text{complete reflection} \quad \text{Eq. (11.6.6)}$$





Tuesday, April 24, 18